

Comprehensive Review and Analysis of SVPWM Technique for CSI and VSI Systems

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Abstract

Power electronics has seen substantial technical development during the last three decades. Recently, it has been applied in a range of industrial, commercial, residential, and aviation applications. Low-power, high-frequency power electronics that must accommodate switching-mode power supplies on demand and high-power, low-frequency power electronics that must accommodate switching-mode power supplies appear to be polarized in opposite ways.

The main goal of this paper is to build SVPWM Model in a systematic manner, use it to compare and contrast Voltage Source Inverters (VSI) and Current Source Inverters (CSI), and apply it to both of these devices. Based on space vector theory, a three-phase VSI and CSI model is developed. The simulation results are found using MATLAB/Simulink for efficiency.

Three-phase inverters are suggested for the SVPWM system. In VSI and CSI modulation, this modulation strategy is essential. We'll focus on meticulously developing the SVPWM Model, using it to compare VSI and CSI on a number of variables, such as load parameters, switching frequency, and modulation index.

Keywords: SVPWM, CSI, VSI, MOSFET, Voltage Source Inverter

1. Introduction

Researchers are putting up their best efforts to maximize energy sources efficiency and meet the demand for electrical power [27]. This includes using the Space Vector Pulse Width Modulation (SVPWM) technology to create or organize converters.

Voltage source inverter (VSI) or current source inverter (CSI) is two main drives. The VSI design has shown to be more beneficial in industrial markets, with the faster dynamic response, improved dependability, and the ability to run motors without de-rating. VSI's fully included intending to save money by increasing efficiency, reducing installation time, and removing the need for connecting power [34].

A wide range of applications is possible because of the fast dynamic comeback for rapid motor torque modification or speed. Minimum components are used to extend the meantime to failure (MTTF), a crucial metric in critical uptime applications [28]. All of these elements combine to create a durable, high-quality industrial design.

In the current environment, the use of force hardware devices in power systems is exceptionally high. Controlling the forced stream in a long-distance transmission line is an urgent need. The powerful drives are delegated either immediate or backhanded change drives—the immediate arrangements with cyclo-converters, while circuitous converter manages VSI and CSI. The VSI is additionally arranged into 2-level High force VSI or Multilevel VSI.

2. Literature Survey

S Pradeepa et. al. (2018) It has been shown that pure air shielding technology (SVPWM) is used to implement and control the converter, and SVPWM pulp is supplied to the converter, and THD analysis is performed on the converter of various details [7]

Shalini et al. (2020) have investigated the Voltage source inverters (VSI) that give desirable inverter output voltages and control the drives in most medium and large power applications. One of the most effective switching control systems used to control inverter outputs is space vector-PWM (SVPWM). An efficient SVPWM approach is modelled and used in two-level Three-Phase VSI in this research. The SVPWM module mainly consists of an Angle generation unit (AGU) using Phase-locked Loop (PLL), Sector generation unit (SGU), time duration module (TDM), Switching time generation (STG) Module, and lastly, SVPWM gate pulse generation (GPG) module [9].



R.Sathishet. al. (2019) Note the use of a quasi-Z SVPWM source inverter for input motor control. The simulation model of the law governing the independent speed of a motor based on the SV -W quasi-Z source inverter. Here, a quasi-Z network is used to amplify the DC input voltage, and the output of a quasi-Z source network is applied to a three-phase inverter based on an SVM. The measuring tube and the frequency AC voltage of the inverter are used to control the speed of the input motor. For the analysis, the simulation results with the R load and the input load are used. It is recommended to use a three-phase LC filter in the output to output the harmonics [15].

M.Venkatesham et. al. (2018) Focus on current PV-based radio / transmission power sources that use surface-to-surface compression technology. Compared to the SPWM method, the power loss of the inverter is reduced by 89%. The conversion loss of the current source inverter is reduced by 65%. In both cases, the amount of energy increases from 2 to 3 times. In addition, it was observed that only one -third of the compression reduction of the latter is used, and output harmonic distortion of the SVPWAM is lower than that of the SPWM. A prototype 1 kW boost- converter-inverter was designed and tested using this model method [16].

The modulation index selected here is the highest possible modulation index, since the largest modulation index is often used in SVPWAM. Theoretically, the THD varies with the index [30]. The DC contact voltage is designed in a manner that it provide continuation of the SVPWM and an six line response ideal envelope of the SVPWAM. Therefore, the harmonics of SVPWAM here do not incorporate the harmonica output from the DC-DC converter. From a mathematical point of view it is a direct comparison of the two methods. The results of the prototype experiment at the laboratory level demonstrate the effectiveness of the proposed strategy in reducing power loss [32].

3. SVPWM for Voltage Source Inverter

Figure 1 depicts a two-level VSI. In this circuit, there are six switches (S1 to S6), and each of these switches is represented by a MOSFET or an IGBT, depending on the nature of the authority function. Unlike other PWM systems, SVPWM makes use of a vector as a reference, which distinguishes it from the others [38].

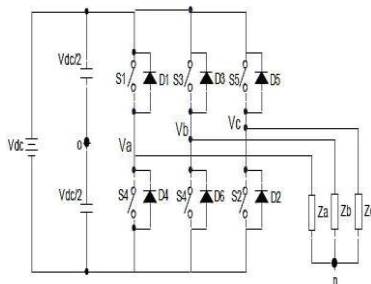


Figure 1: Two Stage Voltage Source Inverter [8]

The switch status of the inverter is shown in Fig 1. If the above switches 1, 3 and 5 are 1, the switch is on or terminal voltage (Va, Vb and Vc) is + Vdc. If upper contact is zero, terminal voltage is zero. The only probable combination of switch modes: 000, 001, 010, 011, 100, 110, 101 and 111, as shown in Figure 1. The ON or OFF states transistors can be used to conclude production power, which is given in Table 1

Table1: Space Vectors, the Switching States and On-State Switches

Type	Switching States	On State Switch
Null Vector	[1 1 1]	s1, s3,s5
	[0 0 0]	s4,s6,s2
	[1 0 0]	s1, s6,s2
Active Vector	[1 1 0]	s1, s3, s2
	[1 1 0]	s4,s3,s2
	[0 1 0]	s4,s3,s2
	[0 1 1]	s4,s3,s5
	[0 0 1]	s4,s3,s5
	[1 0 1]	s1,s6,s5



4. RESULTS AND SIMULATION

Run the simulation and detect waveforms on the Vedic Scope. We will notice that the voltage control inverter performs correctly when a significant imbalance is created at 0.3 sec. Now double-click on the SVPWM generator block and set the proportional gain of the voltage control to zero. The reproduction and note that without voltage control, the DC interface unbalance increments quickly.

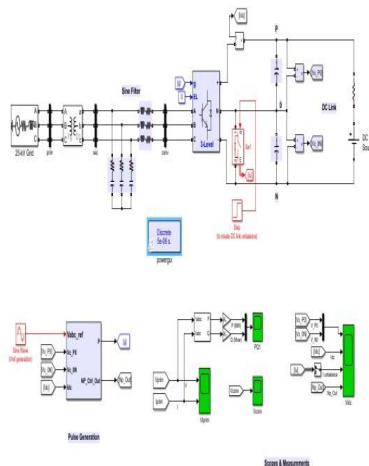


Figure 2: Simulink Model of 3-Level NPC Neutral-Point Inverter Using Space-Vector Pulse-Width-Modulation (SVPWM) Technique with Neutral-Point Voltage Control Inverter

From an ideal 2400-Volt DC source, a 2-MVA, three-phase 3-level inverter delivers power to a 25-kV distribution system. A sine filter is used on secondary side of the distribution transformer to mitigate high-frequency harmonics generated by the inverter. The Sw1 ideal switch is used to create a major unbalance on the DC link.

The inverter is controlled in an open loop using an SVPWM 3-Level Generator block. The PWM switching frequency is set to 1620 Hz, and the neutral-point power control gain is set to 0.04.

5. Simulation Result

Run the simulation and observe waveforms on the VIdc Scope. We will notice that the neutral-point voltage control performs correctly when a significant imbalance is created at 0.2 sec. Now double-click on the Rerun simulation to see how the dc-link imbalance rises quickly without neutral-point voltage management.

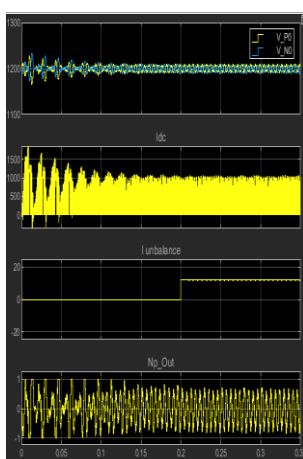


Figure 3: Output Wave Form of Balance and Unbalance Current

The most significant departure of Current in a phase from the mean of three steps is used to compute unbalance.



Furthermore, a can be measured by comparing the intensity of negative sequence currents to positive sequence currents.

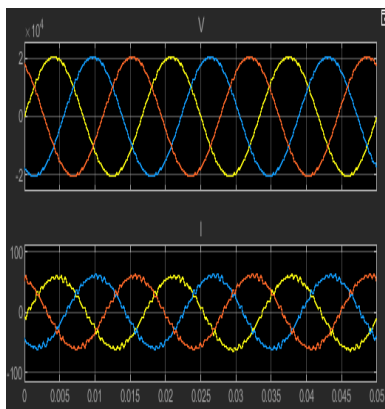


Figure 4: Output Waveform of Neutral-Point Voltage and Current

The scheme phase-to-neutral voltages are used. In general, simply sinusoidal voltages are measured in the formerly reported examination.

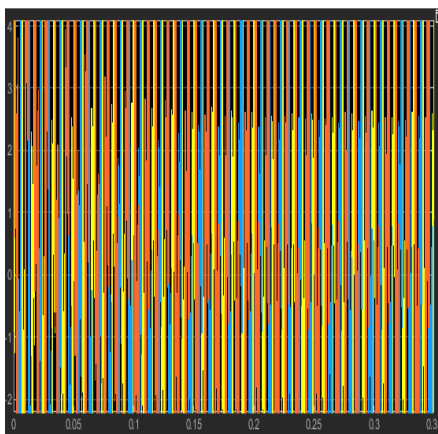


Figure 5: Output Waveform Conjugative Voltage (V_{conj})

The Fast Fourier Transform (FFT) examination of the voltage and present are shown in 6 and 7.

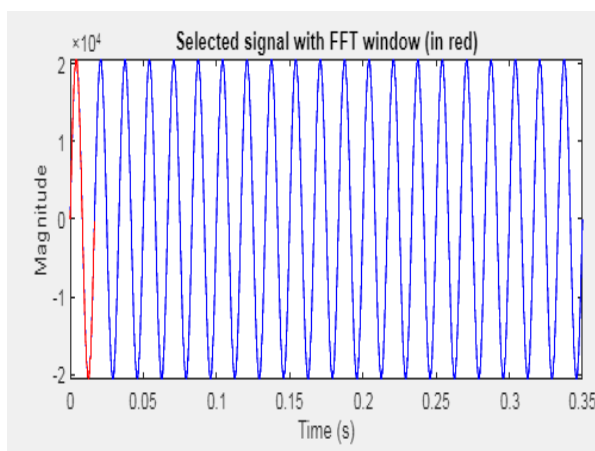


Figure 6: FFT Analysis of Voltage

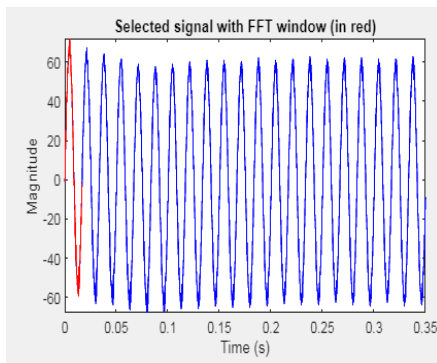


Figure 7: FFT Analysis of Current (I)

The Fast Fourier Transform (FFT) examination of Voltage (V) or THD 1.54 (fundamental frequency 60Hz)

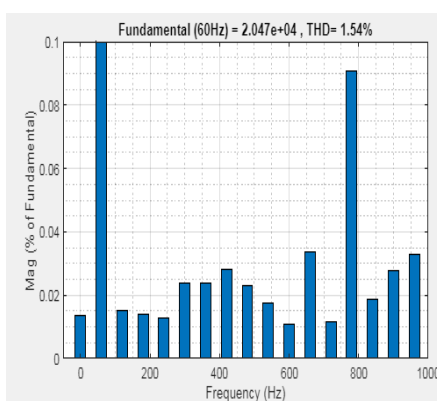


Figure 8: FFT analysis of Voltage (V) and THD 1.54 (fundamental frequency 60Hz)

The Fast Fourier Transform (FFT) analysis of Current (I) and THD 6.30 (fundamental frequency 60Hz)

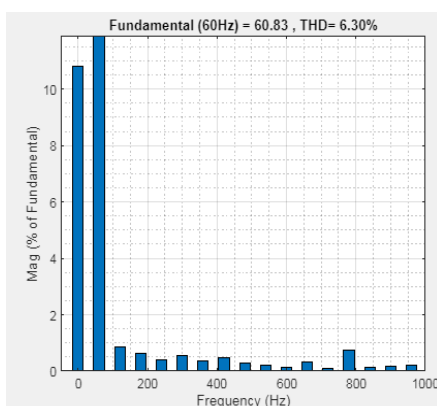


Figure 9: FFT analysis of Current (I) and THD 6.30 (fundamental frequency 60Hz)

Compared to the traditional SPWM and SVPWM methods, the PV SVPWAM control method retains the following advantages.

- 1) Compared to the traditional SPWM inverter system, the switching power loss of VSI is reduced by 89% and CSI is reduced by 65%.
- 2) Due to the reduction of DC capacitors (from 40 to 6 μF) and the need for a small cooling plate, the power density increases 2 times.



3) As the load on passive components, heat sinks and semiconductors is reduced, the cost is reduced by 30%. As shown in the table above, VSI with 6 kHz switching frequency, 560V DC bus voltage and 50 Hz fundamental frequency should provide a smaller load current and load voltage THD compared to the other two VSI specifications. Similarly, a fundamental frequency of 100 Hz in a CSI with a switching frequency of 20 kHz and a DC connection current of 10A compared to the other two CSI specifications should provide less load current and load voltage THD.

Table 2: Overall VSI and CSI With Different Specifications

S.no	Converter	ma	Vdc	Fo (Hz)	FSW (Hz)	cf (µF)	Load		VAb/VI THD %	THD %
							R (Ω)	L (mh)		
1	VSI	0.5	560V	50	6kHz	--	10	50	4.5	0.85
2	VSI	0.5	100 V	100	20kHz	--	10	--	21.03	21.03
3	VSI	0.5	4004.42	50	450Hz	--	20	3	51.43	32
4	CSI	0.5	10A	100	20kHz	30	10	--	1.77	1.89
5	CSI	0.5	30.074	50	6kHz	66	10	50	4.78	3.63
6	CSI	0.5	200A	50	450Hz	66	20	3	5.31	4.34

6. CONCLUSIONS AND FUTURE SCOPE

This article has conducted extensive studies of the modulation strategy of present basis converters and formed a comprehensive theory of permanent or discontinuous carrier-based pulse width methods. In direct and over-regulation, the discussion of these changes and their shortcomings and benefits is emphasized. In the last section, the idea of expanding the meaning of the theory is proposed.

Compare SVPWM of VSI and CSI converter topologies. SVPWM realizes for VSI and CSI, and the algorithm is converse in detail. The focus is on related load, which requires lower THD. For multiple switching frequencies, such as 450 Hz, 6 kHz, and 20 kHz, the operating range of VSI and CSI is displayed. Compare the results of the THD test. By examine overall VSI and CSI models with diverse specifications, the THD% provided by all CSI models is significantly reduced. Therefore, compared to all VSI models, the routine of CSI with SVPWM shows better results.

Later, the demonstrated SVPWM method is planned in the FPGA phase to check the skill of the equipment and planned limits, which are suitable for the use of Hengli gadgets. Further work can recover the efficiency of introduction motor drives by recovering the overall harmonic distortion using SVPWM. The experimental results obtained from the scaling model for the research facility have approved the feasibility of the proposed strategy to reduce troop accidents.

A new hybrid electric vehicle (HEV) uses a buck-boost voltage/power source inverter-based solar and battery using space vector pulse width modulation (SVPWAM) technology to be completed on Matlab. Compared to traditional SPWM and SVPWM methods, the SVPWAM control method retains the following advantages.

1. The shift power loss is reduced by 90% compared to the traditional SPWM inverter system.
2. Due to the reduction of DC capacitors (from 40 to 6 µF) and the need for a small cooling plate, the power density increases twice.
3. As the load on passive components, heat sinks and semiconductors is reduced, the cost is reduced by 30%.

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